

Lecture Notes

11th November 2016

Date FRIDAY / 11th NOV, 16

LECTURE #8

Why do we use Fourier Series?

- F.S is just a means to represent a periodic signal as an infinite sum of sine wave components.
- The main reason to use F.S is that we can better analyse a signal in another domain rather in the original ~~freq.~~ domain.

D.T Fourier Series:-

- When k is changed by any integer multiple of N , the identical sequence is generated.

DETERMINATION OF DISCRETE-TIME FOURIER SERIES COEFFICIENTS:

- We have a sequence $x[n]$ that is periodic with fundamental period N .

$$x[n] = \sum_{k=\langle N \rangle} a_k e^{jk\omega_0 n}$$

- Multiply both sides by $e^{-j\omega_0 mn}$:-

$$e^{-j\omega_0 mn} x[n] = \sum_{k=\langle N \rangle} a_k e^{jk\omega_0 n} e^{-j\omega_0 mn}$$

- Take summation on $n = \langle N \rangle$:

$$\sum_{n=\langle N \rangle} e^{-j\omega_0 mn} x[n] = \sum_{n=\langle N \rangle} \sum_{k=\langle N \rangle} a_k e^{j(k-m)\omega_0 n}$$

$$\sum_{n=\langle N \rangle} e^{-j\omega_0 mn} x[n] = \sum_{k=\langle N \rangle} a_k \sum_{n=\langle N \rangle} e^{j(k-m)\omega_0 n}$$

- Use the sum of geometrical series rule i.e.:

$$\sum_{n=0}^{N-1} a^n = \begin{cases} N & d=1 \\ \frac{1-a^N}{1-a} & d \neq 1 \end{cases}$$

- $\therefore \sum_{n=\langle N \rangle} e^{-j\omega_0 mn} x[n] = \sum_{k=\langle N \rangle} a_k N \delta[k-m] \Rightarrow Na_m$

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$$\rightarrow a_m = \frac{1}{N} \sum_{n=\langle N \rangle} x[n] e^{-j\omega_0 m n}$$

$$\rightarrow a_k = a_{k+N}$$

EXAMPLE # 1:-

$$x[n] = \sin \omega_0 n$$

SOL:-

→ This is the discrete-time counterpart of the signal $x(t) = \sin \omega_0 t$.

→ For the case when $2\pi/\omega_0$ is an integer N , that is when

$$\omega_0 = \frac{2\pi}{N}$$

fundamental

→ $x[n]$ is periodic with period N . and we obtain a result that is exactly analogous to the continuous time case.

→ Expanding the signal as a sum of two complex exponentials, we get.

$$x[n] = \frac{1}{2j} e^{j(2\pi/N)n} - \frac{1}{2j} e^{-j(2\pi/N)n}$$

→ we see by inspection

$$a_1 = \frac{1}{2j}, \quad a_{-1} = -\frac{1}{2j}$$

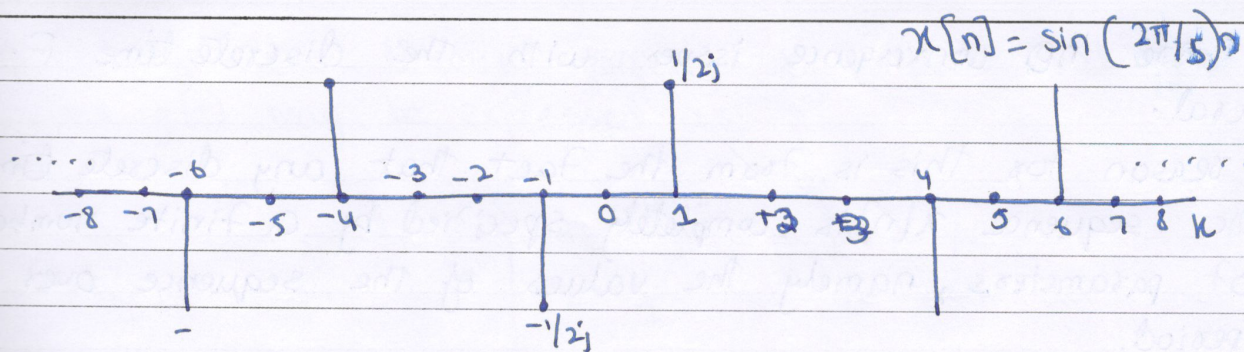
→ remaining coefficients over the interval of summation are zero.

→ As described earlier, these coefficients repeat with period N , thus

$$a_{N+1} = \frac{1}{2j} \text{ and } a_{N-1} = -\frac{1}{2j}$$

→ For example:- F.S coefficients for this example with $N=5$ are shown below.

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→ They repeat periodically.

EXAMPLE #2:-

SOL:-

→ $x[n]=1$ for $-N_1 \leq n \leq N_1$, we choose the length- N interval of summation to include the range $-N_1 \leq n \leq N_1$. The coefficients are given:-

$$a_k = \frac{1}{N} \sum_{n=-N_1}^{N_1} e^{-jk(2\pi/N)n}$$

→ let $m = n + N_1$, then above equation becomes:-

$$a_k = \frac{1}{N} \sum_{n=0}^{2N_1} e^{-jk(2\pi/N)(m-N_1)}$$

$$= \frac{1}{N} e^{jk(2\pi/N)N_1} \sum_{m=0}^{2N_1} e^{-jk(2\pi/N)m}$$

$$a_k = \frac{1}{N} e^{jk(2\pi/N)N_1} \left[\frac{1 - e^{-jk(2\pi/N)(2N_1+1)}}{1 - e^{-jk(2\pi/N)}} \right]$$

$$= \frac{1}{N} \frac{\sin \left[\frac{2\pi k (N_1 + 1/2)}{N} \right]}{\sin(\pi k/N)}, \quad k \neq 0, \pm N, \pm 2N, \dots$$

→ For $k = 0, \pm N, \pm 2N, \dots$, we have

$$a_k = \frac{2N_1 + 1}{N}$$

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→ There are no convergence issues with the discrete time F.S in general.

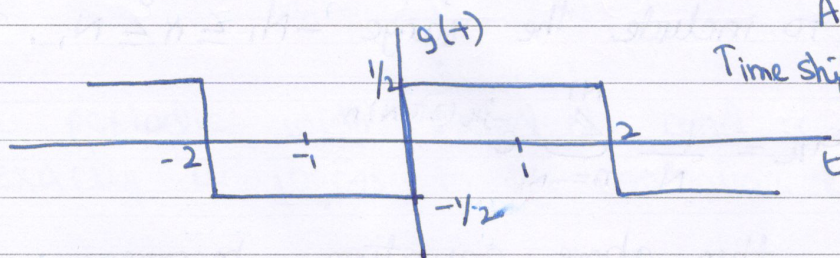
→ The reason for this is from the fact that any discrete time periodic sequence $x[n]$ is completely specified by a finite number N of parameters, namely the values of the sequence over one period.

→ There is no Gibbs phenomenon for DT-Fourier series.

→ The sum of DT FS is finite.

→ No approximation is made so no convergence issue.

EXAMPLE # 3:-



Linearity:-

$$Ax_1(t) + Bx_2(t) \Leftrightarrow Aa_k + Bb_k$$

$$\text{Time shift: } -x(t-t_0) \Leftrightarrow a_k e^{jk\omega_0 t_0}$$

SOL:-

$$T = 4 \quad \& \quad T_1 = 1$$

$$g(t) = x(t-1) - \frac{1}{2} \rightarrow (1)$$

→ According to time shift property, if Fourier series coefficients of $x(t)$ are denoted by a_k , the F.S coefficients of $x(t-1)$ may be expressed as:-

$$b_k = a_k e^{-jk\pi/2}$$

$$\therefore \omega_0 = \frac{2\pi}{T} = \frac{2\pi}{4} \Rightarrow \pi/2$$

→ The Fourier series coefficients of the DC offset in $g(t)$ i.e. the term $-1/2$ on the right hand side of equ (1) are given by:-

$$c_k = \begin{cases} 0 & \text{for } k \neq 0 \\ -1/2 & \text{for } k = 0. \end{cases}$$

→ Applying the linearity property we conclude that the coefficients for $g(t)$ may be expressed as:-

$$d_k = \begin{cases} a_k e^{-jk\pi/2} & \text{for } k \neq 0 \\ a_0 - \frac{1}{2} & \text{for } k = 0 \end{cases}$$

where each a_k may now be replaced by the :-

$$a_k = \frac{\sin(\pi k/2)}{k\pi} e^{jk\pi/2}, \text{ then we have}$$

$$d_k = \begin{cases} \frac{\sin(\pi k/2)}{k\pi} e^{-jk\pi/2} & \text{for } k \neq 0 \\ 0 & \text{for } k = 0 \end{cases}$$

EXAMPLE #4:-

Sol:-

→ Finding the Fourier Series coefficients a_k of the sequence $x[n]$.

→ $N = 5$

→ The signal $x[n]$ may be viewed as the sum of the square wave $x_1[n]$ with Fourier series coefficients b_k and $x_2[n]$ with Fourier series coefficients c_k .

$$a_k = b_k + c_k$$

→ The Fourier series coefficients of $x_1[n]$ is:

$$b_k = \begin{cases} \frac{1}{5} \frac{\sin(3\pi k/5)}{\sin(\pi k/5)} & \text{for } k \neq 0, \pm 5, \pm 10, \dots \\ \frac{3}{5} & \text{for } k = 0, \pm 5, \pm 10, \dots \end{cases}$$

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$$\frac{3}{5} + 1 \Rightarrow \frac{3+5}{5} \Rightarrow 8/5$$

→ The sequence $x_2[n]$ has only a DC value, which is captured by its zeroth Fourier series coefficient.

$$C_0 = \frac{1}{5} \sum_{n=0}^4 x_2[n] = 1$$

→ Since the discrete time Fourier series coefficients are periodic, it follows that $C_k = 1$ whenever k is an integer multiple of 5.

→ The remaining coefficients of $x_2[n]$ must be zero, because $x_2[n]$ contains only a DC component.

→ then,

$$a_k = \begin{cases} b_k = \frac{1}{5} \frac{\sin(k3\pi/5)}{\sin(\pi/5)} & \text{for } k \neq 0, \pm 5, \pm 10, \dots \\ \frac{8}{5} & \text{for } k = 0, \pm 5, \pm 10, \dots \end{cases}$$

$$a_k = \frac{1}{N} \frac{\sin[2\pi k(N+1/2)/N]}{\sin(\pi k/N)}, \quad k \neq 0, \pm 5, \pm 10, \dots$$

