

Signal & Systems

Lecture #9

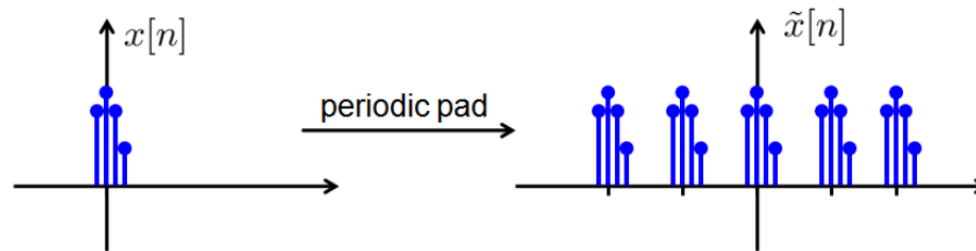
15th May 18

Discrete Time Fourier Transform



Development of the DTFT

- ❖ In deriving discrete-time Fourier Transform we have three key steps:
- ❖ Step#1:
 - ❖ Consider an aperiodic discrete-time signal $x[n]$. We pad $x[n]$ to construct a periodic signal $x'[n]$.



- ❖ Step#2:
 - ❖ Since $x'[n]$ is periodic, by discrete-time Fourier series we have:

$$x'[n] = \sum_{k=\langle N \rangle} a_k e^{jk(2\pi/N)n}$$

Development of the DTFT (cont.)

❖ Where a_k is:

$$a_k = \frac{1}{N} \sum_{n=\langle N \rangle} x'[n] e^{-jk(2\pi/N)n}$$

❖ Here, $\omega_0 = 2\pi/N$.

❖ Now note that $x'[n]$ is a periodic signal with period N and the non-zero entries of $x'[n]$ in a period are the same as the non-zero entries of $x[n]$.

❖ Therefore, it holds that:

$$\begin{aligned} a_k &= \frac{1}{N} \sum_{n=\langle N \rangle} x'[n] e^{-jk(2\pi/N)n} \\ &= \frac{1}{N} \sum_{n=-\infty}^{\infty} x[n] e^{-jk(2\pi/N)n} \end{aligned}$$

Development of the DTFT (cont.)

❖ If we define:

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n}$$

❖ Then:

$$a_k = \frac{1}{N} \sum_{n=-\infty}^{\infty} x[n]e^{-jk(2\pi/N)n} = \frac{1}{N} X(e^{jk\omega_0})$$

❖ Step#3:

❖ Putting above equation in discrete-time Fourier series equation, we have:

$$\begin{aligned} x'[n] &= \sum_{k=\langle N \rangle} a_k e^{jk\omega_0 n} \\ &= \sum_{k=\langle N \rangle} \left[\frac{1}{N} X(e^{jk\omega_0}) \right] e^{jk\omega_0 n} \\ &= \frac{1}{2\pi} \sum_{k=\langle N \rangle} X(e^{jk\omega_0}) e^{jk\omega_0 n} \omega_0, \quad \omega_0 = \frac{2\pi}{N} \end{aligned}$$

Development of the DTFT (cont.)

- ❖ As $N \rightarrow \infty, \omega_0 \rightarrow 0$, so the area becomes infinitesimal small and sum becomes integration and $x'[n]=x[n]$, so above equation becomes,

$$x'[n] = \frac{1}{2\pi} \sum_{k=\langle N \rangle} X(e^{jk\omega_0}) e^{jk\omega_0 n} \omega_0 \rightarrow \frac{1}{2\pi} \int_{2\pi} X(e^{j\omega}) e^{j\omega n} d\omega$$

$$x[n] = \frac{1}{2\pi} \int_{2\pi} X(e^{j\omega}) e^{j\omega n} d\omega$$

- ❖ Hence, the Discrete time Fourier transform pair:

$$x[n] = \frac{1}{2\pi} \int_{2\pi} X(e^{j\omega}) e^{j\omega n} d\omega$$

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}$$

Development of the DTFT (cont.)

- ❖ The first equation is referred to as synthesis equation and second one as analysis equation.
- ❖ $X(e^{j\omega})$ is referred to as the spectrum of $x[n]$.

Example #1

❖ Consider the signal: $x[n] = a^n u[n]$, $|a| < 1$

❖ Solution:

$$\begin{aligned} X(e^{j\omega}) &= \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n} = \sum_{n=-\infty}^{\infty} a^n u[n] e^{-j\omega n} \\ &= \sum_{n=0}^{\infty} (ae^{-j\omega})^n = \frac{1}{1 - ae^{-j\omega}} \end{aligned}$$

Example #2

❖ Consider the signal: $x[n] = a^{|n|}$, $|a| < 1$

❖ Solution:

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} a^{|n|} u[n] e^{-j\omega n} = \sum_{n=-\infty}^{-1} a^{-n} e^{-j\omega n} + \sum_{n=0}^{\infty} a^n e^{-j\omega n}$$

❖ Let $m = -n$ in the first summation we obtain,

$$\begin{aligned} X(e^{j\omega}) &= \sum_{n=-\infty}^{\infty} a^{|n|} u[n] e^{-j\omega n} = \sum_{n=0}^{\infty} a^n e^{-j\omega n} + \sum_{m=1}^{\infty} a^m e^{j\omega m} \\ &= \sum_{n=0}^{\infty} (ae^{-j\omega})^n + \sum_{m=1}^{\infty} (ae^{j\omega})^m \end{aligned}$$

Example #2 (cont.)

- ❖ Both of these summations are infinite geometric series that we can evaluate in closed form, yielding:

$$\begin{aligned} X(e^{j\omega}) &= \frac{1}{1 - ae^{-j\omega}} + \frac{ae^{j\omega}}{1 - ae^{j\omega}} \\ &= \frac{1 - a^2}{1 - 2a\cos\omega + a^2} \end{aligned}$$



The Fourier Transform for Periodic Signals

Periodic Signals

- ❖ For a periodic discrete-time signal: $x[n] = e^{j\omega_0 n}$
- ❖ The discrete-time Fourier transform must be periodic in ω with period 2π .
- ❖ Then the Fourier transform of $x[n]$ should have impulses at ω_0 , $\omega_0 \pm 2\pi$, $\omega_0 \pm 4\pi$, and so on.

- ❖ In fact, the Fourier transform of $x[n]$ is the impulse train:

$$X(e^{j\omega}) = \sum_{l=-\infty}^{\infty} 2\pi\delta(\omega - \omega_0 - 2\pi l)$$

- ❖ Now consider a periodic sequence $x[n]$ with period N and with the Fourier series representation:

Periodic Signals (cont.)

$$x[n] = \sum_{k \in \langle N \rangle} a_k e^{jk(2\pi/N)n}$$

- ❖ In this case, the Fourier transform is:

$$X(e^{j\omega}) = \sum_{k=-\infty}^{\infty} 2\pi a_k \delta\left(\omega - \frac{2\pi k}{N}\right)$$

- ❖ So that the Fourier transform of a periodic signal can directly constructed from its Fourier coefficients.

Example #3

- ❖ Consider the periodic signal:

$$x[n] = \cos \omega_0 n = \frac{1}{2} e^{j\omega_0 n} + \frac{1}{2} e^{-j\omega_0 n}, \text{ where } \omega_0 = \frac{2\pi}{5}$$

- ❖ Solution:

- ❖ From the equation of periodicity we can write:

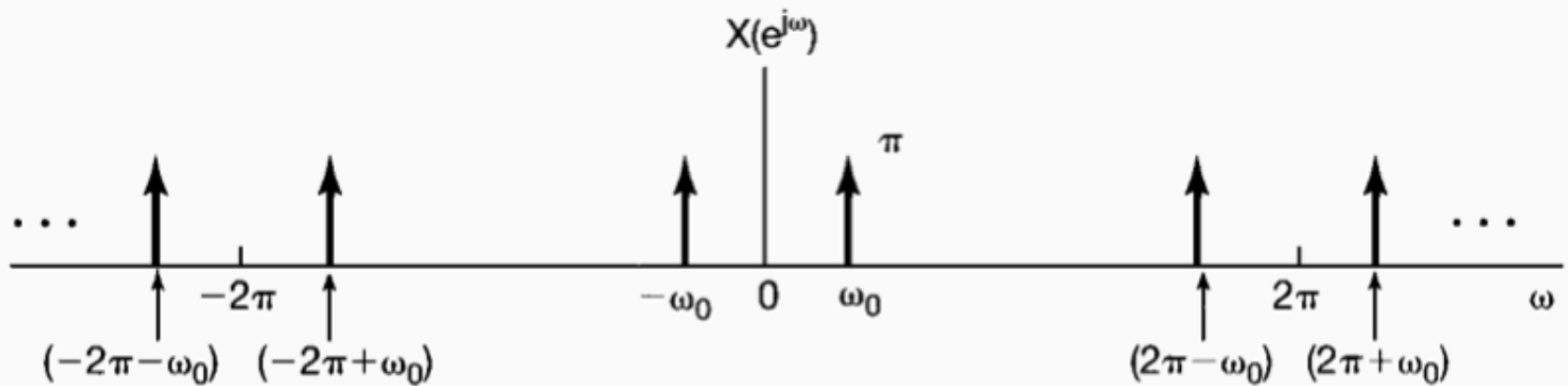
$$X(e^{j\omega}) = \sum_{l=-\infty}^{\infty} \pi \delta\left(\omega - \frac{2\pi}{5} - 2\pi l\right) + \sum_{l=-\infty}^{\infty} \pi \delta\left(\omega + \frac{2\pi}{5} - 2\pi l\right)$$

- ❖ That is,

$$X(e^{j\omega}) = \pi \delta\left(\omega - \frac{2\pi}{5}\right) + \pi \delta\left(\omega + \frac{2\pi}{5}\right), \quad -\pi \leq \omega < \pi$$

- ❖ $X(e^{j\omega})$ repeats periodically with a period of 2π , as shown below:

Example #3 (cont.)



Discrete-time Fourier transform of $x[n] = \cos \omega_0 n$.



Properties of DT Fourier Transform

Periodicity

- ❖ The discrete-time Fourier transform is always periodic in ω with period 2π , i.e.,

$$X\left(e^{j(\omega+2\pi)}\right) = X\left(e^{j\omega}\right)$$

Linearity

❖ If:

$$x_1[n] \leftrightarrow X_1(e^{j\omega})$$

And

$$x_2[n] \leftrightarrow X_2(e^{j\omega})$$

❖ Then:

$$ax_1[n] + bx_2[n] \xleftrightarrow{F} aX_1(e^{j\omega}) + bX_2(e^{j\omega})$$

Time Shifting & Frequency Shifting

❖ If:

$$x[n] \leftrightarrow X(e^{j\omega})$$

❖ Then:

$$x[n - n_0] \stackrel{F}{\leftrightarrow} e^{-j\omega_0 n} X(e^{j\omega})$$

and

$$e^{j\omega_0 n} x[n] \stackrel{F}{\leftrightarrow} X(e^{j(\omega - \omega_0)})$$

Conjugation & Conjugate Symmetry

❖ If: $x[n] \leftrightarrow X(e^{j\omega})$

❖ Then:

$$x^*[n] \overset{F}{\leftrightarrow} X^*(e^{-j\omega})$$

❖ If $x[n]$ is real valued, its transform $X(e^{j\omega})$ is conjugate symmetric. That is:

$$X(e^{j\omega}) = X^*(e^{-j\omega})$$

❖ From this, it follows that $\text{Re}\{X(e^{j\omega})\}$ is an even function of ω and

$\text{Im}\{X(e^{j\omega})\}$ is an odd function of ω .

❖ Similarly the magnitude of $X(e^{j\omega})$ is an even function and the phase angle is an odd function.

Conjugation & Conjugate Symmetry (cont.)

❖ Furthermore,

$$Ev\{x[n]\} \stackrel{F}{\leftrightarrow} \text{Re}\{X(e^{j\omega})\}$$

and

$$Od\{x[n]\} \stackrel{F}{\leftrightarrow} j \text{Im}\{X(e^{j\omega})\}$$

Differencing & Accumulation

❖ If: $x[n] \leftrightarrow X(e^{j\omega})$

❖ Then:

$$x[n] - x[n-1] \xleftrightarrow{F} (1 - e^{-j\omega}) X(e^{j\omega})$$

For signal,

$$y[n] = \sum_{m=-\infty}^n x[m],$$

❖ Its Fourier transform is given as:

$$\sum_{m=-\infty}^n x[m] \xleftrightarrow{F} \frac{1}{(1 - e^{-j\omega})} X(e^{j\omega}) + \pi X(e^{j0}) \sum_{m=-\infty}^{+\infty} \delta(\omega - 2\pi k)$$

Time Reversal

❖ If:

$$x[n] \leftrightarrow X(e^{j\omega})$$

❖ Then:

$$x[-n] \overset{F}{\leftrightarrow} X(-e^{j\omega})$$

Differentiation in Frequency

❖ If:

$$x[n] \leftrightarrow X(e^{j\omega})$$

❖ Then:

$$nx[n] \overset{F}{\leftrightarrow} j \frac{dX(e^{j\omega})}{d\omega}$$

Parseval's Relation

❖ If:

$$x[n] \leftrightarrow X(e^{j\omega})$$

❖ Then:

$$\sum_{n=-\infty}^{+\infty} |x[n]|^2 = \frac{1}{2\pi} \int_{2\pi} |X(e^{j\omega})|^2 d\omega$$

Convolution Property

- ❖ If $x[n]$, $h[n]$ and $y[n]$ are the input, impulse response, and output respectively, of an LTI system, so that,

$$y[n] = x[n] * h[n]$$

then,

$$Y(e^{j\omega}) = X(e^{j\omega})H(e^{j\omega})$$

- ❖ Where $X(e^{j\omega})$, $H(e^{j\omega})$ and $Y(e^{j\omega})$ are the Fourier transforms of $x[n]$, $h[n]$ and $y[n]$ respectively.

Example #4

- ❖ Consider an LTI system with impulse response:

$$h[n] = \delta[n - n_0]$$

- ❖ The frequency response is:

$$H(e^{j\omega}) = \sum_{n=-\infty}^{\infty} \delta[n - n_0] e^{-j\omega n} = e^{-j\omega n_0}$$

- ❖ Thus for any input $x[n]$ with Fourier transform $X(e^{j\omega})$, the Fourier transform of the output is:

$$Y(e^{j\omega}) = e^{-j\omega n_0} X(e^{j\omega})$$

Multiplication Property

❖ It states that:

$$y[n] = x_1[n]x_2[n] \xleftrightarrow{F} Y(e^{j\omega}) = \frac{1}{2\pi} \int_{-\pi}^{\pi} X_1(e^{j\omega}) X_2(e^{j(\omega-\theta)}) d\theta$$

Example #5

- ❖ Consider the signal: $x[n] = \delta[n] + \delta[n-1] + \delta[n+1]$
- ❖ Solution:

$$\begin{aligned} X(e^{j\omega}) &= \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n} \\ &= \sum_{n=-\infty}^{\infty} (\delta[n] + \delta[n-1] + \delta[n+1]) e^{-j\omega n} \\ &= \sum_{n=-\infty}^{\infty} \delta[n] e^{-j\omega n} + \sum_{n=-\infty}^{\infty} \delta[n-1] e^{-j\omega n} + \sum_{n=-\infty}^{\infty} \delta[n+1] e^{-j\omega n} \\ X(e^{j\omega}) &= 1 + e^{-j\omega} + e^{j\omega} = 1 + 2 \cos \omega \end{aligned}$$

Systems Characterized by Linear Constant-Coefficient Difference Equations



Linear Constant-Coefficient Difference Equations

- ❖ A general linear constant-coefficient difference equation for an LTI system with input $x[n]$ and output $y[n]$ is of the form,

$$\sum_{k=0}^N a_k y[n-k] = \sum_{k=0}^M b_k x[n-k]$$

- ❖ Which is usually referred to as Nth-order difference equation.
- ❖ If $x[n] = e^{j\omega n}$ is the input to an LTI system, then the output must be of the form $H(e^{j\omega})e^{j\omega n}$. Substituting these expressions into above equation and performing some algebra allow us to solve for $H(e^{j\omega})$.
- ❖ Based on convolution, above equation can be written as:

$$H(e^{j\omega}) = \frac{Y(e^{j\omega})}{X(e^{j\omega})}$$

Linear Constant-Coefficient Difference Equations (cont.)

- ❖ Applying the Fourier transform to both sides and using the linearity and time-shifting properties we obtain the following expression:

$$\sum_{k=0}^N a_k e^{-jk\omega} Y(e^{j\omega}) = \sum_{k=0}^M b_k e^{-jk\omega} X(e^{j\omega})$$

- ❖ Or equivalently

$$H(e^{j\omega}) = \frac{Y(e^{j\omega})}{X(e^{j\omega})} = \frac{\sum_{k=0}^M b_k e^{-jk\omega}}{\sum_{k=0}^N a_k e^{-jk\omega}}$$

- ❖ The frequency response of the LTI system can be written down by inspection as well.

Example #6

- ❖ Consider the causal LTI system that is characterized by the difference equation:

$$y[n] - ay[n-1] = x[n], \quad |a| < 1$$

- ❖ The frequency response of this system is:

$$H(e^{j\omega}) = \frac{Y(e^{j\omega})}{X(e^{j\omega})} = \frac{1}{1 - ae^{-j\omega}}$$

- ❖ The impulse response is given by:

$$h[n] = a^n u[n]$$

Example #7

- ❖ Consider the LTI system:

$$y[n] - \frac{3}{4}y[n-1] + \frac{1}{8}y[n-2] = 2x[n]$$

- ❖ And let the input to this system be:

$$x[n] = \left(\frac{1}{4}\right)^n u[n]$$

- ❖ Solution:

$$\begin{aligned} Y(e^{j\omega}) &= H(e^{j\omega})X(e^{j\omega}) = \left[\frac{2}{\left(1 - \frac{1}{2}e^{-j\omega}\right)\left(1 - \frac{1}{4}e^{-j\omega}\right)} \right] \left[\frac{1}{1 - \frac{1}{4}e^{-j\omega}} \right] \\ &= \frac{2}{\left(1 - \frac{1}{2}e^{-j\omega}\right)\left(1 - \frac{1}{4}e^{-j\omega}\right)^2} \end{aligned}$$

Example #7 (cont.)

- Using the partial fraction expansion, we get:

$$Y(e^{j\omega}) = \frac{B_{11}}{1 - \frac{1}{4}e^{-j\omega}} + \frac{B_{12}}{\left(1 - \frac{1}{4}e^{-j\omega}\right)^2} + \frac{B_{21}}{1 - \frac{1}{2}e^{-j\omega}}$$

- Solving the partial fraction gives:

$$B_{11} = -4, \quad B_{12} = -2, \quad B_{21} = 8$$

- So that:

$$Y(e^{j\omega}) = -\frac{4}{1 - \frac{1}{4}e^{-j\omega}} - \frac{2}{\left(1 - \frac{1}{4}e^{-j\omega}\right)^2} + \frac{8}{1 - \frac{1}{2}e^{-j\omega}}$$

Example #7 (cont.)

❖ The inverse transform i.e., $y[n]$ is:

$$y[n] = \left\{ -4\left(\frac{1}{4}\right)^n - 2(n+1)\left(\frac{1}{4}\right)^n + 8\left(\frac{1}{2}\right)^n \right\} u[n]$$



Problems

Problem #1

- ❖ Use the Fourier transform analysis equation to calculate the Fourier transforms of the following signals:

$$(a): \left(\frac{1}{2}\right)^{|n-1|}$$

$$(b): \delta[n+2] - \delta[n-2]$$

$$(c): \sin\left(\frac{\pi}{2}n\right) + \cos(n)$$

Problem #2

- ❖ Consider a causal and stable LTI system S whose input $x[n]$ and output $y[n]$ are related through the second-order difference equation:

$$y[n] - \frac{1}{6}y[n-1] - \frac{1}{6}y[n-2] = x[n]$$

- ❖ (a): Determine the frequency response $H(e^{j\omega})$ for the system S .
- ❖ (b): Determine the impulse response $h[n]$ for the system S .



Thank You!