

Signal & Systems

Lecture # 8 Fourier Series-II

20th December 18

Discrete Time Periodic Signals

Fourier Series Representation of DT

- The Fourier series representation of discrete-time periodic signal is finite as opposed to the infinite series representation required for continuous-time periodic signals.

Linear Combinations of Harmonically Related Complex Exponentials

- A discrete-time signal $x[n]$ is periodic with period N if: $x[n] = x[n+N]$.
- The fundamental period is the smallest positive N and the fundamental frequency is $\omega_0 = \frac{2\pi}{N}$.
- The set of all discrete-time complex exponential signals that are periodic with period N is given by:

$$\phi_k[n] = e^{jk\omega_0 n} = e^{jk(2\pi/N)n}, \quad k = 0, \pm 1, \pm 2, \dots$$

- All of these signals have fundamental frequencies that are multiples of $2\pi/N$ and thus are harmonically related.

Linear Combinations of Harmonically Related Complex Exponentials (cont.)

- There are only N distinct signals in the set this is because the discrete-time complex exponentials which differ in frequency by a multiple of 2π are identical. That is: $\phi_k[n] = \phi_{k+rN}[n]$

- The representation of periodic sequences in terms of linear combinations of the sequences $\Phi_k[n]$ is:

$$x[n] = \sum_k a_k \phi_k[n] = \sum_k a_k e^{jk\omega_0 n} = \sum_k a_k e^{jk(2\pi/N)n}$$

- Since the sequences $\Phi_k[n]$ are distinct over a range of N successive values of k , the summation in above equation need include terms over this range.

Linear Combinations of Harmonically Related Complex Exponentials (cont.)

- Thus the summation is on k as k varies over a range of N successive integers beginning with any value of k .
- We indicate this by expressing the limits of the summation as $k=\langle N \rangle$. That is:

$$x[n] = \sum_{k=\langle N \rangle} a_k \phi_k[n] = \sum_{k=\langle N \rangle} a_k e^{jk\omega_0 n} = \sum_{k=\langle N \rangle} a_k e^{jk(2\pi/N)n}$$

Discrete-Time Fourier Series Coefficients

- Assuming $x[n]$ is square-summable i.e., $\sum_{n=-\infty}^{\infty} |x[n]|^2 < \infty$ or $x[n]$ satisfies the Dirichlet conditions.

- In this case we have:

$$x[n] = \sum_{k=\langle N \rangle} a_k e^{jk\omega_0 n} = \sum_{k=\langle N \rangle} a_k e^{jk(2\pi/N)n}, \quad \text{Synthesis Equation}$$

$$a_k = \frac{1}{N} \sum_{n=\langle N \rangle} x[n] e^{-jk\omega_0 n} = \frac{1}{N} \sum_{n=\langle N \rangle} x[n] e^{-jk(2\pi/N)n}, \quad \text{Analysis Equation}$$

- As in continuous time, the discrete-time Fourier series coefficient a_k are often referred to as the spectral coefficients of $x[n]$.
- These coefficients specify a decomposition of $x[n]$ into a sum of N harmonically related complex exponentials.

Example #1

- Consider the signal:

$$x[n] = \sin \omega_0 n$$

- Which is the discrete-time counterpart of the signal.
- $x[n]$ is periodic only if $2\pi/\omega_0$ is an integer or a ratio of integers.

Example #2

- The signal $x[n] = \sin(2\pi n/3)$ is periodic with fundamental period $N_0=3$. calculate the DTFS coefficients.

Properties of Fourier Series Coefficients

Linearity

- For continuous-time Fourier series, we have:

$$x_1(t) \Leftrightarrow a_k \quad \text{and} \quad x_2(t) \Leftrightarrow b_k$$

$$Ax_1(t) + Bx_2(t) \Leftrightarrow Aa_k + Bb_k$$

- For Discrete-time case, we have:

$$x_1(t) \Leftrightarrow a_k \quad \text{and} \quad x_2(t) \Leftrightarrow b_k$$

$$Ax_1(t) + Bx_2(t) \Leftrightarrow Aa_k + Bb_k$$

Time Shift

$$x(t - t_0) \leftrightarrow a_k e^{-jk\omega_0 t_0}$$

$$x[n - n_0] \leftrightarrow a_k e^{-jk\Omega_0 n_0}$$

- Proof: Let us consider the Fourier series coefficient b_k of the signal $y(t) = x(t - t_0)$. $b_k = \frac{1}{T} \int_T x(t - t_0) e^{-jk\omega_0 t} dt$
- Letting $\tau = t - t_0$ in the integral, we obtain:

$$\frac{1}{T} \int_T x(\tau) e^{-jk\omega_0(\tau + t_0)} dt = e^{-jk\omega_0 t_0} \frac{1}{T} \int_T x(\tau) e^{-jk\omega_0 \tau} dt$$

where $x(t) \leftrightarrow a_k$. Therefore,

$$x(t - t_0) \leftrightarrow a_k e^{-jk\omega_0 t_0}$$

Time Reversal

$$x(-t) \leftrightarrow a_{-k}$$

$$x[-n] \leftrightarrow a_{-k}$$

- Proof: Consider a signal $y(t) = x(-t)$. The Fourier series representation of $x(-t)$ is:

$$x(-t) = \sum_{-\infty}^{\infty} a_k e^{-jk2\pi t/T}$$

- Letting $k = -m$, we have:

- Thus:
$$y(t) = x(-t) = \sum_{m=-\infty}^{\infty} a_{-m} e^{jm2\pi t/T}$$

$$x(-t) \leftrightarrow a_{-k}$$

Time Scaling

- Time scaling is an operation that in general changes the period of the underlying signal.
- Specifically if $x(t)$ is periodic with period T and fundamental frequency $\omega_0 = 2\pi/T$, then $x(\alpha t)$, where α is a positive real number, is periodic with period T/α and fundamental frequency $\alpha\omega_0$.

Properties of Continuous-Time Fourier Series

- **Multiplication:**

- Suppose that $x(t)$ and $y(t)$ are both periodic with period T and that:

$$x(t) \leftrightarrow a_k$$

$$y(t) \leftrightarrow b_k$$

- Since the product $x(t)y(t)$ is also periodic with period T , we can expand it in a Fourier series with Fourier series coefficients h_k expressed in terms of those for $x(t)$ and $y(t)$. The result is:

$$x(t)y(t) \leftrightarrow h_k = \sum_{l=-\infty}^{\infty} a_l b_{k-l}$$

- The sum on the R.H.S may be interpreted as the Discrete-time convolution of the sequence $x(t)$ and $y(t)$.

Properties of Continuous-Time Fourier Series (cont.)

- Conjugation and Conjugate Symmetry:

- Real $x(t) \leftrightarrow a_{-k} = a_k^*$ (conjugate symmetric)

- Real & Even $x(t) \leftrightarrow a_k = a_k^*$ (real and even a_k)

- Real & Odd $x(t) \leftrightarrow a_k = -a_k^*$ (purely imaginary and odd a_k),
 $a_0 = 0$

- Even part of $x(t) \leftrightarrow \text{Re}\{a_k\}$

- Odd part of $x(t) \leftrightarrow j\text{Im}\{a_k\}$

Properties of Continuous-Time Fourier Series (cont.)

- Parseval's Relation:

- Parseval's relation for continuous-time periodic signal is:

$$\frac{1}{T} \int_T |x(t)|^2 dt = \sum_{k=-\infty}^{\infty} |a_k|^2$$

- Where a_k are the Fourier series coefficients of $x(t)$ and T is the period of the signal.
- L.H.S of the above equation is the average power (i.e., energy per unit time) in one period of the periodic signal $x(t)$.

- Also:

$$\frac{1}{T} \int_T |a_k e^{jk\omega_0 t}|^2 dt = \frac{1}{T} \int_T |a_k|^2 dt = |a_k|^2$$

- So that $|a_k|^2$ is the average power in the k th harmonic component of $x(t)$.
- Thus Parseval's relation states that the total average power in a periodic signals equals the sum of the average powers in all of its harmonic components.

Properties of Discrete-Time Fourier Series

- Multiplication:

- In discrete-time, suppose that: $x[n] \Leftrightarrow a_k$
and
 $y[n] \Leftrightarrow b_k$

- Are both periodic with period N . then the product $x[n]y[n]$ is also periodic with period N .
- Its Fourier coefficients d_k are given by:

$$x[n]y[n] \Leftrightarrow d_k = \sum_{l=\langle N \rangle} a_l b_{k-l}$$

- The result is a periodic convolution between the FS sequences.
- $w[n]$ is periodic with N .

Properties of Discrete-Time Fourier Series (cont.)

- First Difference:

- If $x[n]$ is periodic with period N , then so is $y[n]$, since shifting $x[n]$ or linearly combining $x[n]$ with another periodic signal whose period is N always results in a periodic signal with period N .

- Also, if:
$$x[n] \leftrightarrow a_k$$

- Then the Fourier coefficients corresponding to the first difference of $x[n]$ may be expressed as:

$$x[n] - x[n-1] \leftrightarrow \left(1 - e^{-jk(2\pi/N)}\right) a_k$$

Properties of Discrete-Time Fourier Series (cont.)

- Parseval's Relation:

- Parseval's relation for discrete-time periodic signals is given by:

$$\frac{1}{N} \sum_{n=\langle N \rangle} |x[n]|^2 = \sum_{n=\langle N \rangle} |a_k|^2$$

- The average power in a periodic signal = the sum of the average power in all of its harmonic components.

Fourier Series & LTI Systems

Fourier Series & LTI Systems

- The response of a continuous-time LTI system with impulse response $h(t)$ to a complex exponential signal e^{st} is the same complex exponential multiplied by a complex gain:

$$y(t) = H(s)e^{st}$$

where

$$H(s) = \int_{-\infty}^{\infty} h(\tau)e^{-s\tau} d\tau$$

- In particular, for $s=j\omega$, the output is $y(t)=H(j\omega)e^{j\omega t}$.
- The complex functions $H(s)$ and $H(j\omega)$ are called the system function (or transfer function) and the frequency response, respectively.

Fourier Series & LTI Systems (cont.)

- By superposition, the output of an LTI system to a periodic signal represented by a Fourier series:

$$x(t) = \sum_{k=-\infty}^{\infty} a_k e^{jk\omega_0 t} = \sum_{k=-\infty}^{\infty} a_k e^{jk(2\pi/T)t} \quad \text{is given by}$$

$$y(t) = \sum_{k=-\infty}^{\infty} a_k H(jk\omega_0) e^{jk\omega_0 t}$$

- That is, the Fourier series coefficients b_k of the periodic output $y(t)$ are given by:

$$b_k = a_k H(jk\omega_0)$$

- Similarly, for discrete time signals and systems, response $h[n]$ to a complex exponential signal $e^{j\omega n}$ is the same complex exponential multiplied by a complex gain:

Fourier Series & LTI Systems (cont.)

$$y[n] = H(jk\omega_0) e^{jk\omega_0 n}$$

where

$$H(e^{j\omega}) = \sum_{n=-\infty}^{\infty} h[n] e^{-j\omega n}$$

Example #3

- Suppose that the periodic signal $x(t) = \sum_{k=-3}^3 a_k e^{jk2\pi t}$ with $a_0=1$, $a_1=a_{-1}=1/4$, $a_2=a_{-2}=1/2$, and $a_3=a_{-3}=1/3$ is the input signal to an LTI system with impulse response $h(t)=e^{-t}u(t)$.

The End
